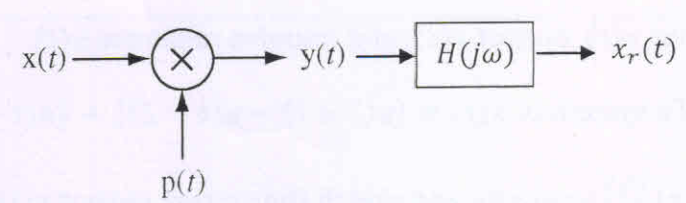


Q1	<p>A causal LTI system is described by the difference equation</p> $y[n] = y[n - 1] + 0.5y[n - 2] + x[n] + x[n - 1], \text{ obtain}$ <p>a) Impulse response $h[n]$, and sketch pole-zero plot of $H(z)$</p> <p>b) Step response $s[n]$, and sketch pole-zero plot of $S(z)$</p> <p style="text-align: right;">[8+7]</p>	[15]
Q2	<p>The signal $y(t) = e^{-2t} u(t)$ is the output of a causal all-pass system for which the system function is</p> $H(s) = \frac{s - 1}{s + 1}$ <p>(a) Find two possible inputs $x(t)$ that could produce $y(t)$ and sketch respective ROC's</p> <p>(b) What is the input $x(t)$ if it is known that $\int_{-\infty}^{\infty} x(t) dt < \infty$</p> <p>(c) Find a stable (but not necessarily causal) impulse response $h_1(t)$ of a system that has $x(t)$ as an output and $y(t)$ is the input.</p> <p style="text-align: right;">[8+2+5]</p>	[15]
Q3	<p>An LTI system has input $x(t)$, output $y(t)$, and impulse response $h(t)$.</p> <p>If the step response of a system is $s(t) = \{u(2 + t) - u(t - 2)\} * \{u(t + 2)u(2 - t)\}$,</p> <p>and input $x(t) = \frac{d^2}{dt^2} \left\{ s\left(\frac{t}{2}\right) \right\}$, compute and sketch the system output $y(t)$,</p> <p>where $*$ represents linear convolution.</p>	[15]
Q4	<p>A periodic signal $x[n]$ has period 8 and Fourier coefficients a_k. Sketch one period of $x[n]$, with the help of the following information</p> <p>i. $a_k = -a_{k-4}$.</p> <p>ii. $x[2n + 1] = (-1)^n$.</p> <p>Now, if a periodic signal $y[n]$ with period 8 and Fourier coefficients b_k is represented as below, obtain b_k in terms of a_k, and sketch one period of $y[n]$</p> $y[n] = \left(\frac{1 + (-1)^n}{2} \right) x[n - 1]$ <p style="text-align: right;">[8+7]</p>	[15]

<p>Q5</p>	<p>A discrete-time LTI system has input $x[n]$, impulse response $h[n]$, and output $y[n]$.</p> <p>Impulse response $h[n]$ and input $x[n]$ are represented below</p> $h[n] = 8(-1)^n \left(\frac{\sin\left(\frac{\pi n}{8}\right)}{\pi n} \cdot \frac{\sin\left(\frac{3\pi n}{8}\right)}{\pi n} \right)$ $x[n] = \frac{16}{\pi} x_1[n] \cos\left(\frac{3\pi n}{8}\right) + 2x_2[n] \cos\left(\frac{7\pi n}{8}\right)$ <p>where</p> $x_1[n] \xleftrightarrow{DTFT} X_1(e^{j\omega}) = \begin{cases} j\omega & \omega < \frac{\pi}{8} \\ 0 & \omega > \frac{\pi}{8} \end{cases} \quad x_2[n] \xleftrightarrow{DTFT} X_2(e^{j\omega}) = \begin{cases} \cos(4\omega) & \omega < \frac{\pi}{8} \\ 0 & \omega > \frac{\pi}{8} \end{cases}$ <p>Sketch $X(e^{j\omega})$, $H(e^{j\omega})$, and $Y(e^{j\omega})$ for the interval $-2\pi \leq \omega \leq 2\pi$. [10+5+5]</p>	<p>[20]</p>
<p>Q6</p>	<p>A sampling process is represented by the figure below:</p>  <p>$x(t) = (\cos(100t))^2$ is an input signal, $p(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT_s)$ is an impulse train, n is an integer, T_s is sampling period, $h(t) = \left(\frac{\sin 200t}{\pi t}\right)^2$ is the impulse response of a filter.</p> <p>a. Sketch the magnitude response of $X(j\omega)$ and $Y(j\omega)$, if the sampling frequency is 25% above the Nyquist rate.</p> <p>b. If the sampling frequency is 25% below the Nyquist rate, sketch the magnitude response of $X_r(j\omega)$ and $H(j\omega)$.</p> <p>Represent the reconstructed signal $x_r(t)$. [10+10]</p>	<p>[20]</p>