

# Comprehensive Exam (Part–II)

Instructor-in-charge: B. Sainath, 2210-A, Electrical and Electronics Eng. Dept., BITS Pilani.

Course No./Title : EEE F430/Green communications and Networks

DATE: May 15<sup>th</sup>, 2023. Test duration: 180–(Part–I) Mins. Max. Points: 25.

**Note:** Answer the following questions. Provide key intermediate steps. ( $5 \times 5 = 25$  points)

**Q. 1.** Figure 1 depicts the wireless power transfer (WPT) circuit based on inductive coupling. Recall the Tesla strategy. Suppose that for an incremental subinterval  $\Delta\omega$ ,  $|I_\ell(\Delta\omega)| \approx |I_s(\Delta\omega)|$ . Let  $L_2 = \beta M$ ,  $0 < \beta < 1$ ,  $r_\ell C_\ell = \tau_\ell$ . Answer the following:

(i) Obtain a quadratic equation in  $M$  in the form  $\mathcal{A}M^2 + \mathcal{B}M + \mathcal{C} = 0$ . Explicitly write the real factors  $\mathcal{A}$ ,  $\mathcal{B}$ , and  $\mathcal{C}$ , which depend on the WPT circuit parameters. Determine the roots.

(ii) Determine the conditions which guarantee that out of the two real roots, one is negative and one is positive.

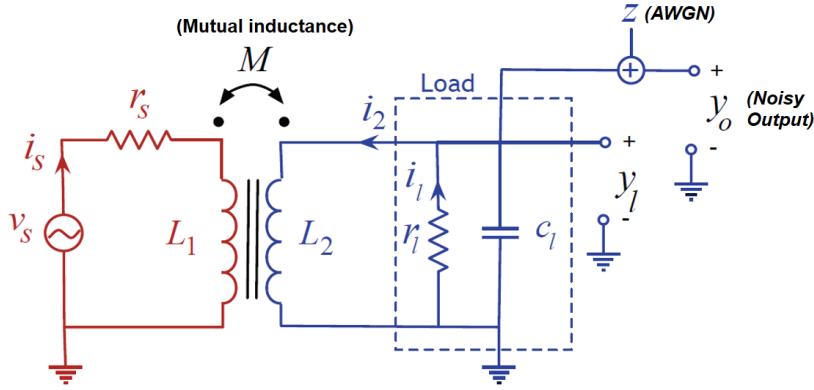


Fig. 1: Pertaining to Q.1.

(iii) Assume that, for the measured subinterval  $\Delta\omega$ ,  $|I_\ell(\Delta\omega)|^2 \approx |I_s(\Delta\omega)|^2$ . Obtain an approximate expression for the ratio of  $r_\ell$  and  $r_s$  in terms of  $\eta(\Delta\omega)$ . For  $\eta(\Delta\omega) = 0.9$ , determine  $r_\ell$  when  $r_s = 1 \text{ k}\Omega$ .

[3 + 1 + 1 points]

**Q. 2.** Consider the fast fading wireless channel scenario. The optimization problem is defined as follows.

$$\max_{P_1, P_2, \dots, P_L \geq 0} \frac{1}{L} \sum_{\ell=1}^L \log_2 \left( 1 + \frac{P_\ell \gamma_\ell}{\sigma_n^2 + \sigma_i^2} \right),$$

such that  $\frac{1}{L} \sum_{\ell=1}^L P_\ell = \bar{P}$ , for all  $\gamma_\ell$ .

Use the following notation.

- $L$  denotes the number of parallel subchannels that fade independently.
- $P_\ell$ ,  $\ell = 1 \leq \ell \leq L$  denotes the power allocation for  $\ell^{\text{th}}$  subchannel.
- $\gamma_\ell \geq 0$  denotes  $\ell^{\text{th}}$  subchannel power gain.
- $\sigma_n^2$  denotes the white Gaussian noise variance and  $\sigma_i^2$  denotes the interference variance.
- $\bar{P}$  is the average power constraint.

Answer the following: (i) Let  $P_\ell^*$  is the optimal power allocation for  $\ell^{\text{th}}$  subchannel. Using the Lagrangian approach, obtain the water-filling solution in terms of  $\lambda^* > 0$ ,  $\gamma_\ell$ , for  $\sigma_n^2 = 0 \text{ dB}$  and  $\sigma_i^2 = 0 \text{ dB}$ . How can you determine  $\lambda^*$ ? Comment. Write down the average power constraint using  $P_\ell^*$  as  $L \rightarrow \infty$ .

(ii) Suppose the transmitter has knowledge of CSI (i.e.  $\gamma$ ) and transmits with adaptive power  $P_t$ . Write down the expressions for fading-averaged spectral efficiency (FASE) and fading-averaged energy efficiency (FAEE). Note: Assume that  $P_{est}$  denotes the power consumed for acquiring CSI. Further, assume that  $P_{crt}$  denotes the power consumed by the circuitry.

[3 + 2 points]

**Q. 3.** Consider a dual hop, TS–SWIPT cooperative wireless system as shown in the Fig. 2. In it, all are single antenna nodes, and the SWIPT relay is half-duplex. The source transmits at a fixed power  $P_s$ . Without loss of generality, let the noise variance at the relay node is unity. Let the frequency-flat fading channel power gain  $\gamma_{sr} \triangleq |h|^2 \sim \exp(1)$  and the first hop path loss factor is denoted by  $L_p^{S \rightarrow R}$ . Let  $\Gamma_R$  denote the instantaneous SNR at the EH relay receiver, and the conversion efficiency is  $\eta$ .

